

Article

A Photovoltaic-fed Z-source Inverter Motor Drive with Fault-tolerant Capability for Rural Irrigation

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Abstract: In recent years, photovoltaic (PV) systems have emerged as economical solutions for irrigation systems in rural areas. However, they are characterized by low voltage output and less reliable configurations. To address this issue in this paper, a promising inverter configuration called Impedance (Z)-source inverter (ZSI) is designed and implemented to obtain high voltage output with single-stage power conversion, particularly suitable for irrigation application. An improved and efficient modulation scheme and design specifications of the network parameters are derived. Additionally, a suitable fault-tolerant strategy is developed and implemented to improve reliability and efficiency. It incorporates an additional redundant leg with an improved control strategy to facilitate the fault-tolerant operation. The proposed fault-tolerant circuit is designed to handle switch failures of the inverter modules due to the open-circuit and short-circuit faults. The relevant simulation and experimental results under normal, faulty and post-fault operation are presented. The post-fault operation characteristics are identical to the normal operation. The motor performance characteristics such as load current, torque, harmonic spectrum, and efficiency are thoroughly analysed to prove the suitability of the proposed system for irrigation applications. This study provides an efficient and economical solution for rural irrigation utilized in developing countries, for example, India.

Keywords: Fault-tolerant; open-circuit fault; short-circuit fault; induction motor drive; Z-source network; inverter; Photovolataic (PV), IGBT.

1. Introduction

India is an agriculture-based developing country with its major economic contribution from the agriculture sector. Around 70% of the farming area is dependent on irrigation in rural areas. It is because monsoon-based cultivation cannot cater the huge demand. According to the All India Crop Estimation Survey Report 2013-14, above 85% of the farming land is irrigated by the ground water irrigation system; and the rest is dependent on surface-water irrigation system. Generally, sources of electric power and diesel engines are used for irrigation to drive motor pumps, but these sources of energy are still not sufficient to fulfill the demand. With unlimited usage of motor drives, the energy demand is increasing worldwide leading to a burden on conventional sources of energy. However, the depletion of conventional sources of energy has attracted power generation from solar and wind [1–3]. Of these, solar energy is considered to be more effective as harnessing of solar is much easier than wind power. With limited access to electricity in rural areas, solar-powered irrigation systems may serve as an alternative for irrigation in off-grid areas. Additionally, the Ministry of New and Renewable Energy (MNRE), Government of India supports installation of solar systems in India with attractive incentives, so that PV modules are now affordable with the relevant subsidies.

34 But, the low output voltage of PV systems is a matter of concern for motor-pump applications. To
35 cater for this difficulty, a PV input source is fed through configurations of DC-DC converter to step up
36 the low voltage of the PV output. Various configurations of PV-fed dc-dc converters were discussed in
37 [4]. This paper [4] presented a comparative analysis of cascaded converter topologies for PV systems.
38 The outputs of these converters are fed through inverters to ac motor drives utilized in pumping
39 applications. Nevertheless, these converters are not sufficient to cater to the higher voltage output
40 requirement of motor drive applications. Moreover, single-stage power conversion is not supported by
41 the conventional converter-inverter topologies which is an essential aspect of economical design. This
42 problems can be overcome by impedance-source inverters (abbreviated as Z-source inverter, ZSI) [5–8].
43 It provides extended output-voltage range, lower harmonic distortion, and high power factor. With the
44 aforementioned advantages, the ZSI module is desirable for PV fed motor-drive systems for pumping
45 applications. Studies in [9–11] show the behavior analysis of impedance networks for PV systems.
46 The Z-source inverter topology is suitable to achieve maximum power point tracking in applications
47 utilizing photovoltaics without affecting the gain of Z-source inverter operation. The suitability of
48 the Z-source inverter is also explored in detail in [12]. The proposed control in [12] ensures the
49 simultaneous operation of the given topology integrated with the three-Level Neutral Point Clamped
50 (NPC) Inverter in buck-boost mode and induction motor control. Also, several topologies of Z-source
51 inverter are proved to be suitable for high gain output requirements such as electric vehicles [13] and
52 induction motor drives [14]. The converter possesses a good tracking of the reference voltage and
53 disturbance-rejection characteristics.

54 However, the operation of these drives is prone to get affected by potentially high risks factors.
55 Once such factor, as suggested by the results from industry statistics, is switch faults leading to
56 permanent failures of power switching devices. The most common switch faults in inverter modules
57 are open-circuit and short-circuit faults. The adverse effects of switch faults on motor drive systems
58 were thoroughly discussed in [15]. Thus, the reliability of operation for such drives is questionable
59 under the absence of an efficient fault-tolerant scheme. The need of fault-tolerance becomes more
60 prominent in the case of renewable-sources-fed electric-motor drives. Various modified topologies are
61 suggested depending on the nature of the applications. Some of the fault-tolerant strategies are feasible
62 for open-circuit or short-circuit faults, and the literature also discussed other topologies for multiple
63 faults. In [16–19] the fault-tolerant strategies towards these faults are discussed for conventional
64 converter circuits. These topologies provide wider control for fault detection and diagnosis. A
65 few more studies [20–29] suggested protection methods and fault diagnosis techniques for switch
66 faults. However, the characteristics like complexity of control schemes, high cost of operation and
67 longer detection time makes them unsuitable for irrigation application. A comprehensive review
68 of fault-tolerant techniques was presented in [30–34]. Among the suggested methods, addition of a
69 redundant leg enables efficient control during fault-tolerant operation. The above literature review
70 suggested that the motor-driven solar water pumping system should be fed through an high-gain
71 economical power conversion arrangement to reduce cost of the system. In addition to it, the proposed
72 system should be equipped with an effective fault-tolerant scheme which should compensate for
73 switch failures occurred in the inverter module. This arrangement eliminates the risk of reduced
74 volume of water during long-hours operation for irrigation in rural areas.

75 The major contributions of this paper are summarized as follows:

- 76 1. Design and implementation of a complete efficient irrigation system using a high gain single-stage
77 power conversion, Z-source inverter (ZSI) and a PV system. With the proposed modulation
78 strategy, the ZSI can use higher modulation index with lower voltage stress across the switching
79 devices as compared to the existed noncoupled impedance networks.
- 80 2. Develop a fast fault-tolerant control algorithm and control measures to facilitate continuous
81 operation during the event of a fault. The fault-diagnosis is achieved effectively within 20 ms
82 while maintaining nearly rated power during post-fault operation.

83 3. Analyze the performance parameters of the proposed system to verify its compatibility for
84 irrigation applications.

85 In light of the above, an efficient motor-drive system with fault-tolerant capability is proposed
86 in this paper. With operation at high modulation index, the proposed system has improved power
87 quality. Also, it proves to be an economical solution with less devices required. The economical feature
88 makes this topology affordable to users in rural areas. The response of fault-tolerant control schemes
89 under open-circuit and short-circuit faults is investigated for irrigation application. The fault diagnosis
90 is performed very quickly. In addition to it, due to presence of PV source with battery storage systems,
91 the pumping operation is independent of frequent power outage and variations in weather conditions.

92 The remainder of this paper is organized into the following sections. The design of the PV array,
93 Z-source network and motor specifications for water-pump are described in Section 2. Section 3
94 describes the proposed fault-tolerant strategy for the system under the faulty modes. The methods of
95 fault detection and flow of fault compensation control scheme are thoroughly explained in this section.
96 The relevant simulation results under normal, faulty and post fault operation are analyzed in Section 4.
97 Experimental results are presented in Section 5 which were obtained for a test bench of 1-kW induction
98 motor prototype interfaced through dSPACE. The suitability of the proposed system for the irrigation
99 application is investigated in Section 6. The obtained results shows the validity and effectiveness of
100 the developed system with fault-tolerant capability for irrigation applications. Finally, a conclusion is
101 presented in Section 7.

102 **2. System Design Specifications**

103 In the conventional inverter-drive system, the induction-motor drive fed by a PV source operates
104 with two stages of power conversion: the first stage involves conversion of the low DC output of the
105 photovoltaic array to a constant DC by a high gain boost converter, and the second stage incorporates
106 the conversion of a constant DC voltage to a variable voltage and variable frequency supply using
107 VSIs. In this paper, the two-stage power conversion is replaced by a single-stage conversion with the
108 ZSI topology. It consists of two inductances and two capacitances connected in the X-shape to facilitate
109 the power flow. The circuit diagram of the proposed system is shown in Figure 1. It consists of the PV
110 source, Z-source inverter for high output voltage gain, and motor-pump arrangement for irrigation
111 application. The PV source acts as input supply for the inverter module. The ZSI network is responsible
112 for supplying the ac power to the motor-pump arrangement. The switching modulation strategies of
113 the inverter module, discussed in later part of the paper, are chosen such that the maximum output is
114 obtained across the load.

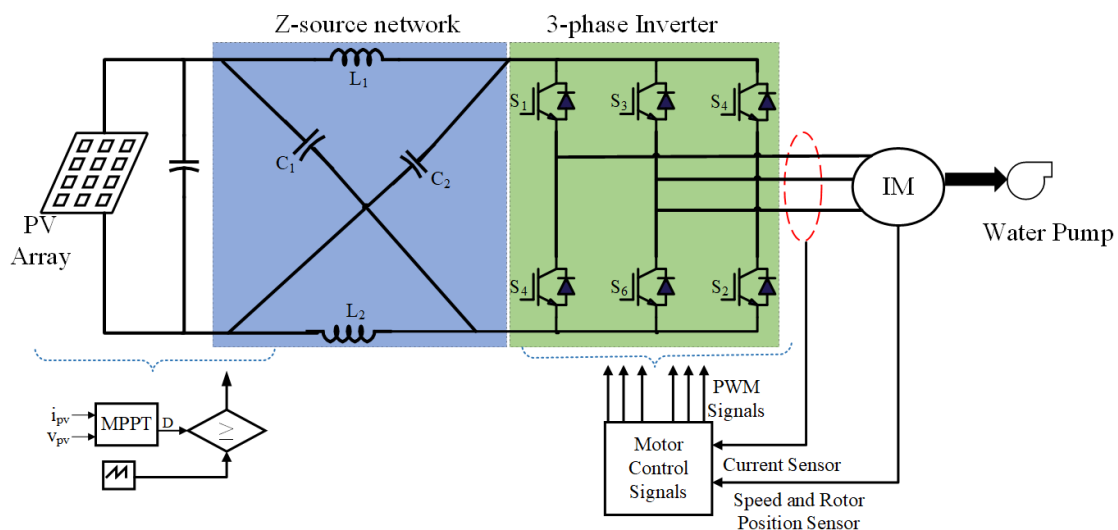


Figure 1. Circuit diagram of the PV-fed ZSI-fed motor drive for pumping application.

115 2.1. Selection of PV Array Parameters

PV solar cells have nonlinear current-voltage (I-V) characteristics. The output voltage and power vary with the irradiance and temperature. The operating point is derived from the intersection of the load line with the PV voltage-current characteristic [35–38]. The PV source was modeled using the characteristic equation:

$$i_{pv} = i_L - i_o \left(e^{\frac{q(v_{pv} + i_{pv} R_s)}{\eta k T}} - 1 \right) \quad (1)$$

116 where i_{pv} is the PV cell current, v_{pv} is the cell voltage, i_L is the photo-current, η is quality factor, i_o is
 117 diode current, k is Boltzmann constant, q is the charge of an electron, R_s is the cell resistance, and T is
 118 the operating temperature in Kelvin.

119 The detailed design parameters of the Solar PV array are given in Table 1.

Table 1. Design parameters of Solar PV Array.

Symbol	Description	Value
V_o	Open-circuit voltage of one module	21.6 V
I_o	Short-circuit current of one module	0.64 A
V_m	MPP voltage	600 V
I_m	MPP current	14.5 A
N_s	Number of series-connected modules	34
N_p	Number of parallel-connected modules	25

120 2.2. Z-source network

121 In the proposed modulation scheme, as shown in Figure 2, the active state is kept intact. It is
 122 achieved by implementing a shoot-through state while operating the power semiconductor switch
 123 in zero-state. This results in high voltage gain due to independent control operation of the active
 124 and shoot-through states. The shoot-through state is applied across all the three arms of the inverter
 125 module to minimize the stress on the power switches. The ZSI proves to be suitable for a PV system
 126 using the implemented modulation technique. The proposed modulation strategy provides advantage
 127 of current sharing and thereby makes the circuit economical, with a requirement of low current rating
 128 devices.

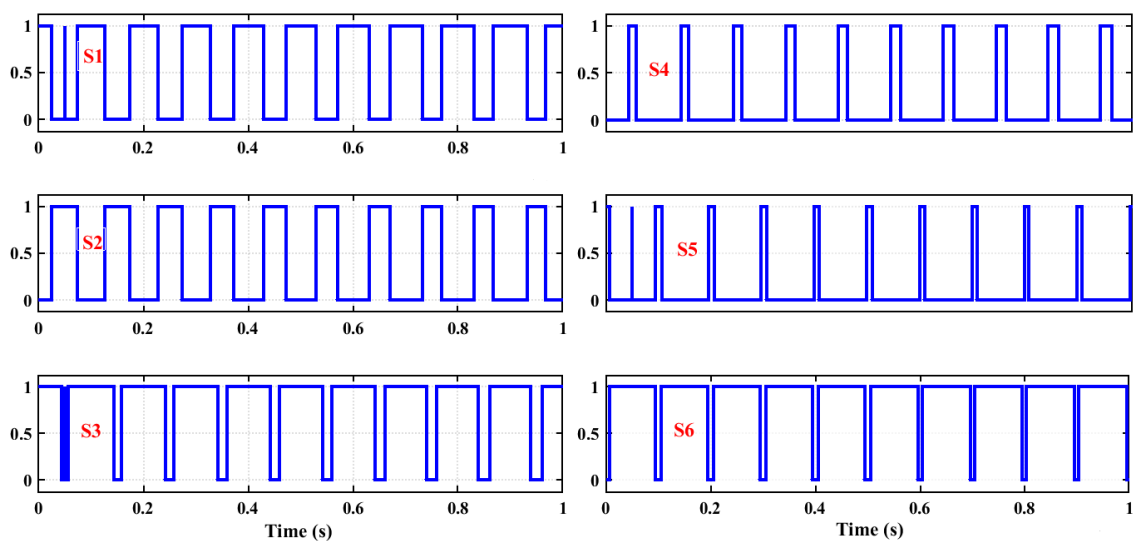


Figure 2. Proposed modulation scheme.

129 In addition to efficient modulation scheme, the selection of inductances and capacitances is vital
 130 for single-stage conversion to obtain high voltage gain.

The PV dc voltage and output ac peak voltage are related as [39]:

$$v_o = \frac{M \cdot B \cdot V_{pv}}{2} \quad (2)$$

131 where v_o is the output peak voltage, B is the boost factor, M is the modulation index, and V_{pv} is the PV
132 output voltage.

Modulation index is calculated as:

$$M = \frac{V_s}{V_t} \quad (3)$$

133 where, V_s is the amplitude of sinusoidal reference waveform and V_t is the amplitude of triangular
134 waveform.

The boost factor (B) is expressed as:

$$B = \frac{1}{2M - 1} \quad (4)$$

The inductor value of the Z-source network is calculated by

$$L = \frac{V_c \cdot D}{2f_s K_L I_{LM}} \quad (5)$$

135 where K_L is the inductor-current ripple. For high gain operation, it should be $\leq 2\%$.

The capacitor connected in the impedance source network can be calculated as:

$$C = \frac{I_{LM} \cdot D}{f_s K_c V_{link}} \quad (6)$$

136 where K_c is the measure of ripple coefficient of the voltage across capacitor, V_{link} is the dc link voltage,
137 I_{LM} is the peak value of the inductor current and f_s is the switching frequency.

138 From equations (2-6), the calculated parameters of the Z-source network are given in Table 2.

Table 2. Parameters of Z-source network.

Symbol	Description	Specification
M	Modulation Index	0.72
B	Boost factor	1.92
$L_1 = L_2$	Inductance	1.3 mH
$C_1 = C_2$	Capacitance	1000 μ F
D	Duty cycle	0.5
f_c	Carrier frequency	10 kHz
f_{st}	Shoot-through frequency	20 kHz

139 2.3. Specifications of Motor-Pump Arrangement

An appropriate design of the induction motor plays a important role in the proposed application of the water-pump arrangement. In order to have sufficient supply of water, the nonlinear relationship between load torque and motor speed is considered. It is given by

$$T_L = K\omega_m^2 \quad (7)$$

140 where T_L is load torque, K is the constant of proportionality of the pump, ω_m is the speed of the motor.

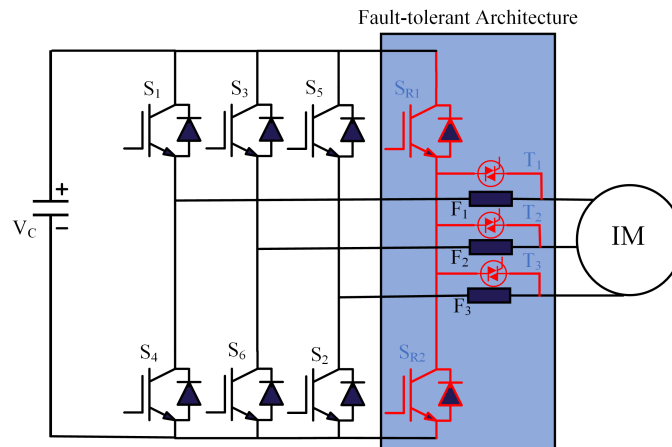
141 The specifications of the motor are listed in Table 3.

Table 3. Motor Specifications.

Symbol	Description	Specification
P	Number of poles	6
P_m	Power	4 kW
N	Speed	1500 rpm
R_s	Stator Resistance	0.6 Ω
R_r	Rotor Resistance	0.6 Ω
J	Moment of Inertia	0.05 $kg.m^2$
L_m	Mutual Inductance	0.03 H

142 3. Proposed Fault-tolerant strategy

143 In this paper, the term "fault-tolerant" depicts that the system will continue to operate in a normal
 144 satisfactory manner even after the occurrence of fault. The proposed architecture of the utilized
 145 fault-tolerant strategy is shown in Figure 3.

**Figure 3.** Proposed fault-tolerant strategy.

146 The proposed fault-tolerant topology is suitable for switch failures due to open-circuit and
 147 short-circuit faults in the inverter module. The proposed fault-tolerant strategy includes:

- 148 (a) Fault detection and localization
- 149 (b) Fault-compensation for post-fault operation

150 It incorporates an additional redundant leg to facilitate the fault tolerant operation. With regard
 151 to the PV-source input, the redundant fault-tolerant topology is utilized using external thyristors
 152 to facilitate the transfer of power flow from the faulty leg to the redundant leg. Upon detection of
 153 open-circuit fault, the gate signals of the faulty legs are disconnected and the redundant leg is activated.
 154 In case of a short-circuit fault, the high-rated fuses safeguard the motor terminals and the the current
 155 falls to zero to prevent the motor from being damaged. The redundant leg is normally unused and
 156 comes to operation only during the occurrence of the fault.

157 3.1. Fault Detection and Localization

158 In order to activate the proposed fault-tolerant strategy, the nature of fault is required to be
 159 identified. The fault detection is performed by the measurement of variations in output-current value
 160 or the voltage amplitude. Several detection methods for open-circuit and short-circuit faults are
 161 suggested in literature [40–42]. Upon detection, the location of the fault should be identified. The
 162 capability of a system to identify the location of a fault quickly adds to the speedy recovery of the
 163 system from the events of the fault.

164 3.1.1. Open-circuit fault detection

In the faulty mode, huge oscillations are introduced in the q-axis current which induce high torque ripples in the motor drive. Various detection methods for open-circuit faults are suggested in the literature. Based on the ease of implementation on the experimental prototype, the method of slope measurement of the trajectory of space vector is utilized to detect the open-circuit fault. It involves calculation of slope for the trajectory of space vector during operation based on current measurement of d-q axis (i_{dk}, i_{qk}). The slope is defined as:

$$y = \frac{i_{dk} - i_{d(k-1)}}{i_{qk} - i_{q(k-1)}} \quad (8)$$

165 In this method, the faulty phase is identified using the measured slope from eq (8). In order
 166 to identify the faulty switch in the faulty phase, the polarity of the phase current is examined. A
 167 positive faulty phase current indicates the fault in the lower-switch and a negative faulty phase current
 168 indicates the fault in the upper-switch of the faulty leg. The above-mentioned faulty leg and switch
 169 detection method is shown in Table 4.

Table 4. Detection of Faulty Phase/Switch.

Faulty Switch	Slope	Current in Faulty Phase
S_1	0	-
S_2	0	+
S_3	$\sqrt{3}$	-
S_4	$\sqrt{3}$	+
S_5	$-\sqrt{3}$	-
S_6	$-\sqrt{3}$	+

170 3.1.2. Short-circuit fault detection

171 Due to the catastrophic consequences of short-circuit faults, the fault detection and isolation of the
 172 faulty leg must be quick. In this paper, the short-circuit fault is detected by gate-voltage compensation
 173 method. In this method, the gate voltage is compared with a reference voltage. The circuit diagram of
 174 the proposed method for short-circuit fault detection is shown in Figure 4. The fault detection circuit
 175 consists of a differential generator in series with a short-circuit fault detector. It measures the difference
 176 between the gate voltage and the reference voltage. In case of a faulty condition, the fault current is
 177 reduced by increasing the voltage across the Zener diode. This method is very fast and can be easily
 178 integrated to a small IC, which results in lesser size and economical solution to build the experimental
 179 prototype.

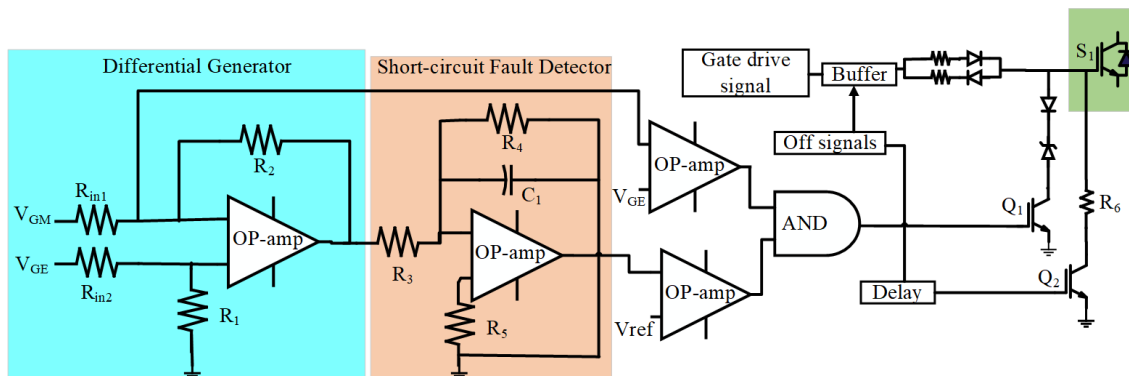


Figure 4. Short-circuit fault detection circuit.

180 3.2. Fault Compensation

181 Once the fault is detected, the fault compensation control scheme should come into the action.
 182 The fault compensation signifies that the diagnosis algorithm should be designed in such a way that it
 183 brings the system back to the normal operation instead of no-operation in event of fault. In case of
 184 the open-circuit faults, the control circuit sends the command and the faulted leg is transferred to the
 185 redundant leg (S_{R1} and S_{R2}) with corresponding external thyristors (T_1, T_2, T_3). The fault detection
 186 is fast and the motor receives normal current. Similarly, once the short-circuit fault is detected,
 187 the hardware protection circuit stops the gate signal to all semiconductor switches and protective
 188 high-rating fuses are inserted in the system. The gate signals are transferred to the new leg and motor
 189 continues to run the same as in case of pre-fault conditions. The control scheme to implement the
 190 above-mentioned fault-tolerant strategy is shown in Figure 5.

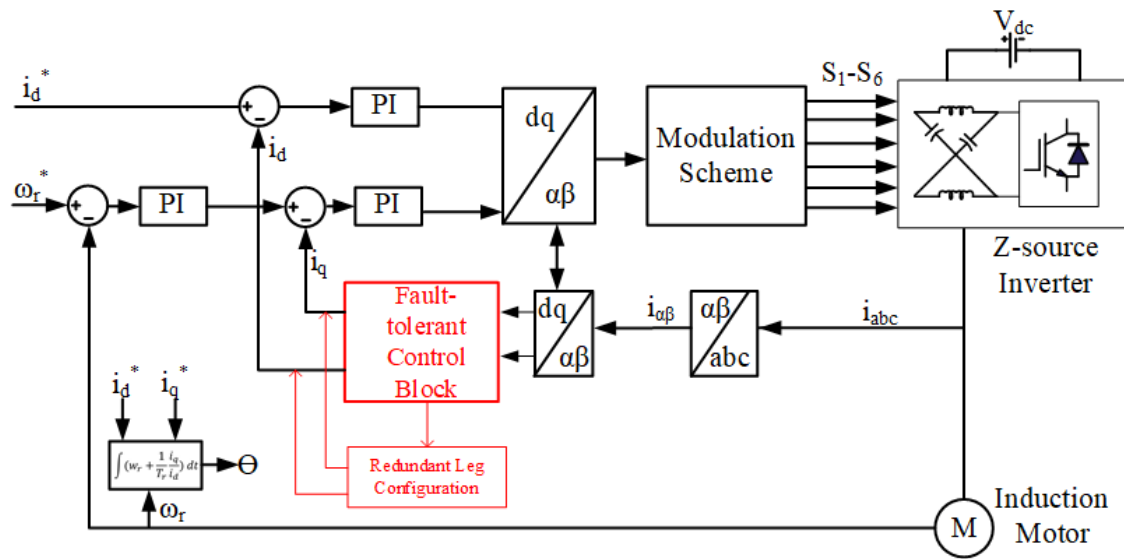


Figure 5. Control scheme of the proposed fault-tolerant strategy.

The proposed fault-tolerant strategy consists of an modulation scheme arrangement, inner-ring current controller, outer-ring speed controller, phase current measurement module of inverter, fault-detection and compensation circuit, and the redundant leg. The d - q -axis current of the three-phase winding of the motor under normal operation is expressed as follows:

$$\begin{bmatrix} i_d \\ i_q \end{bmatrix} = \begin{bmatrix} \cos\theta & \sin\theta \\ -\sin\theta & \cos\theta \end{bmatrix} \times \begin{bmatrix} i_a - \frac{(i_b + i_c)}{2} \\ \frac{\sqrt{3}(i_a + ib)}{2} \end{bmatrix}$$

When the phase winding suffers from an open-circuit fault, the current of the respective phase is reduced to zero. In case of fault in phase-A of the winding, the d - q -axis current changes and is given as:

$$\begin{bmatrix} i_d \\ i_q \end{bmatrix} = \begin{bmatrix} \cos\theta & \sin\theta \\ -\sin\theta & \cos\theta \end{bmatrix} \times \begin{bmatrix} -\frac{(i_b + i_c)}{2} \\ \frac{\sqrt{3}(i_a + ib)}{2} \end{bmatrix}$$

191 In the event of open-circuit fault, the motor works in an asymmetric state and the magnetic field
 192 will no longer be round rotating magnetic field. It results in drop in the output voltage. In case of
 193 motor-drive applications, the fault-tolerant strategy is labelled efficient if it is capable of maintaining
 194 the air-gap magnetic field to be round rotating magnetic field. The fault-tolerant control should ensure
 195 that voltage vectors are kept constant during the normal and post-fault operation. If a short circuit
 196 fault occurs in the switches of phase leg-A, the high rating fuses must blow to prevent the propagation
 197 of fault in the system. The fuses are selected based on integral current square rating $\int i^2 t$ of the devices.

198 It should be lower than the rated value of power semiconductor switch. Upon detection of fault,
 199 the gating signals must be transferred to the redundant arm. For desired operation of motor-pump
 200 arrangement, the dip in speed is also considered. The motor speed measurement is performed and the
 201 controllers are tuned in such a way that sufficient current is supplied to the faulty-phase in event of
 202 the fault. The switch over to the redundant circuit must be quick.

203 A flowchart depicting the structure of the proposed fault-tolerant control scheme is shown in
 204 Figure 6. The fault detection is relatively fast and ensures immunity from the false alarms.

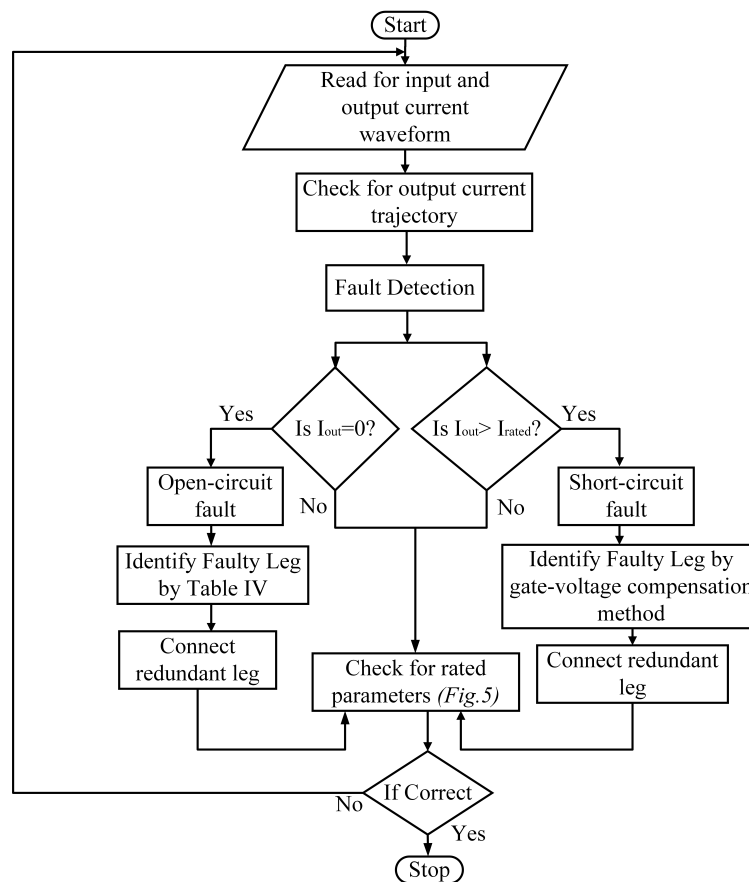


Figure 6. Flowchart of the proposed fault-tolerant strategy.

205 4. Results and Discussions

206 Several case studies have been conducted using the MATLAB/Simulink to investigate the
 207 performance of the proposed system under normal, faulty and post-fault operation. The system
 208 parameters are defined in Section 2 of this paper. The relevant case studies are discussed as follows:

209 4.1. Normal Operation

210 The objective of this case study is to verify the operation of the proposed system under normal
 211 operation. The output current waveform under the no-fault operation is shown in Figure 7a. Under
 212 normal operation, the two capacitors limit the voltage and current ripple on the inverter module to
 213 provide a sinusoidal output. The maximum capacitor voltage rating is equal to twice the peak voltage
 214 of the energy-storage device. There is no flow of discontinuous current through the inductors and
 215 capacitors. This results in continuous conduction mode (CCM) operation with a high reactive power.
 216 It can be inferred that the ZSI supplies a less distorted output current, as shown in Figure 7a, to the
 217 motor-pump arrangement.

218 4.2. Faulty Operation

219 The propagation of faults from faulty areas to normal areas results in damage to the system
 220 components. It may lead to a complete shutdown of the system. In an electrical system, transients are
 221 introduced in the event of a fault which should be avoided. In this part of the study, the variation in
 222 output current of the system under the switch failures due to open-circuit and short-circuit fault are
 223 presented.

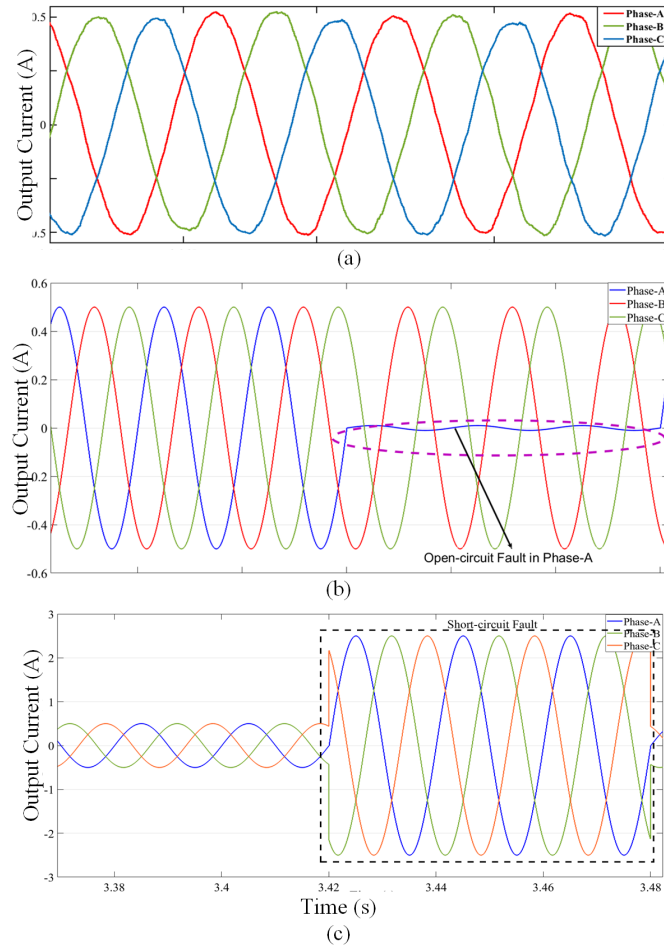


Figure 7. Simulation results of output current: (a) normal operation, (b) open-circuit fault, and (c) short-circuit fault.

224 4.2.1. Case 1: Open-circuit switch fault

225 An open-circuit fault occurs due to the thermal cycling. For the simulation studies, the open-circuit
 226 fault is implemented in the IGBTs of phase-A. An open-circuit fault is created by turning off the gate
 227 signal of the power semiconductor switch. The output current waveform during this fault is shown
 228 in Figure 7b. It is observed that due to the switch failures occurred due to open-circuit fault, the
 229 corresponding phase current is zero.

230 4.2.2. Case 2: Short-circuit switch fault

231 An IGBT short-circuit fault may occur due to high/wrong gate voltage, due to driver circuit
 232 failure, power supply failure or due to $\frac{dv}{dt}$ disturbance, and temperature rise or overvoltage stress.
 233 Short-circuit faults are difficult to deal with, as the time interval between the fault occurrence and
 234 diagnosis is very small. The faulty mode induces a nonzero dc component current in the stator winding.
 235 This results in dynamic braking of the motor drive. There are high chances of device failures in this

236 faulty condition. The motor drive is not suitable to operate in this condition. The output current
 237 waveform is shown in Figure 7c. It can be seen from Figure 7c that there is an high rise in the output
 238 current across the faulty phase. This high current will damage the circuit components and hence is
 239 required to be prevented from flowing in the system.

240 4.2.3. Case 3: Motor performance

241 The motor-pump arrangement is the essential part of the irrigation application. The performance
 242 of the motor is required to be investigated under faulty operation to show the adverse effects of
 243 switch failure in the given system. In this aspect, the variations in torque and speed of the motor are
 244 studied. It can be observed from Figure 8a that under the faulty operation, the electromagnetic torque
 245 is increased due to the fault. The torque tends to show oscillations with high magnitudes which can
 246 lead to torsional vibrations and shaft failures. The operation of the motor drive is not suitable in such
 247 condition. In addition to it, there is a sudden fall in speed of the motor due to fault and hence motor
 248 eventually stops as shown in Figure 8b.

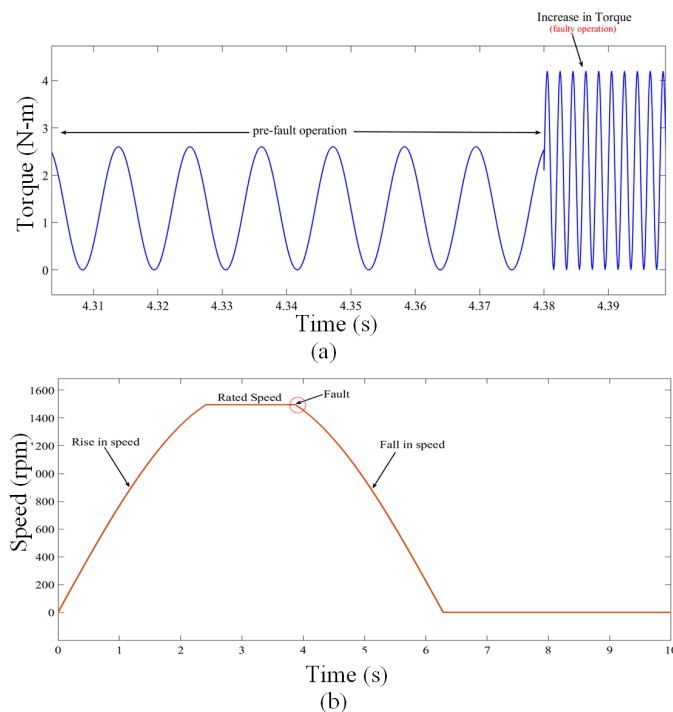


Figure 8. Motor performance under faulty operation: (a) torque-response, (b) speed response.

249 From the above analyses, it is clear that the conventional system without fault-tolerant capability
 250 is not suitable for irrigation application.

251 4.3. Post-Fault Operation

252 Based on the above analysis, it is recommended to implement an efficient fault-tolerant strategy
 253 in order to prevent propagation of the faults in the system. The output current waveform for the
 254 fault-tolerant control (FTC) operation under open-circuit fault and short-circuit are shown in Figure 9.
 255 The fault-tolerant operation is shown with a series of events; (i) pre-fault operation, (ii) faulty operation
 256 and (iii) post-fault operation after the implementation of the proposed fault-tolerant strategy.

257 It can be clearly seen in Figure 9a and 9b that the faults are introduced at $t = 4.38$ s and due to
 258 proposed fault-tolerant strategy the faults are cleared at an interval of 20 ms. Upon implementation of
 259 fault-tolerant strategy, the output current characteristics are identical to pre-fault condition. Moreover,
 260 the torque is restored back to the nominal value by utilizing the proposed fault-tolerant scheme. It

261 can be observed in Figure 9c that torque increases for the fault duration and restores back to the rated
 262 value during fault-tolerant operation. The diagnosis time can be reduced by changing the switching
 263 sequence of the redundant circuit.

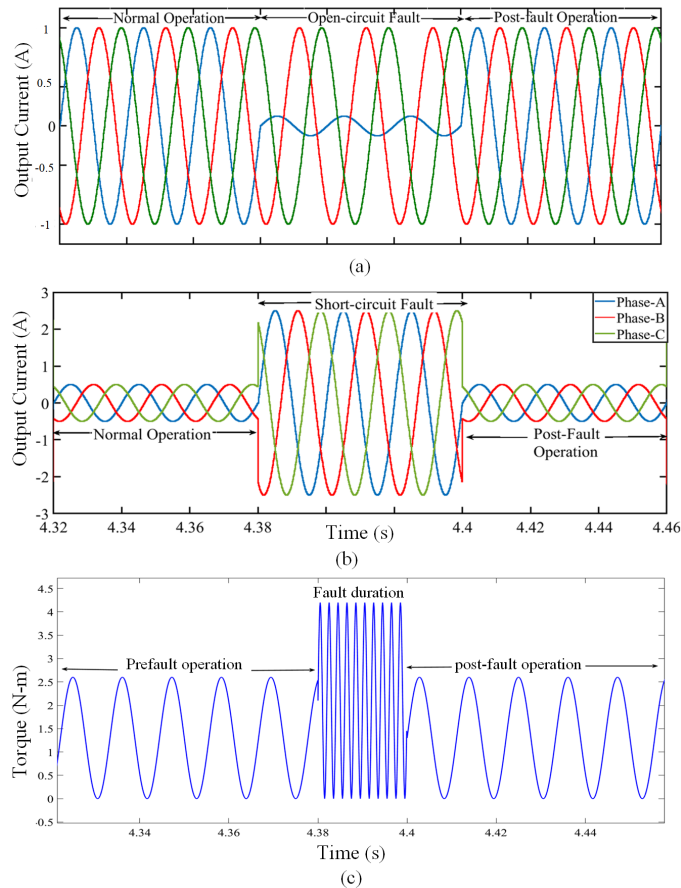


Figure 9. Simulation results under FTC operation: (a) output current under open-circuit fault, (b) output current under short-circuit fault, and (c) torque response.

264 The main advantage of using the redundant configuration for the proposed system is that the
 265 system is able to maintain the dc-link voltage balance under normal and faulty conditions. Due to
 266 the proposed strategy, the output current waveform is ripple-free. The post-fault current and torque
 267 characteristics are identical to the normal operation characteristics. Also, the flyback effect is taken
 268 into consideration while implementing the switching sequence. This improves the reliability factor
 269 of the proposed technique for the motor-drive system. The primary goal of utilizing every voltage
 270 output of the source is fulfilled in this fault-tolerant strategy in order to provide the maximum output
 271 to the motor drive for irrigation applications.

272 5. Experimental Results

273 A laboratory prototype has been realized for the implementation of the proposed drive system
 274 under normal, faulty and fault-tolerant operation in this section. The experiments are conducted on a
 275 1kW motor test bed, as shown in Figure 10, in order to validate the theoretical analysis and signify the
 276 performance of the proposed fault diagnosis algorithm. The test bench is operated at an input voltage
 277 of 100 V and output voltage of 300 V. The fault injection and control algorithm are interfaced through
 278 dSPACE control desk software. The fault incidents are manually injected in the drive system. The
 279 faults are introduced in phase-A IGBTs of the inverter module for the experimental studies.

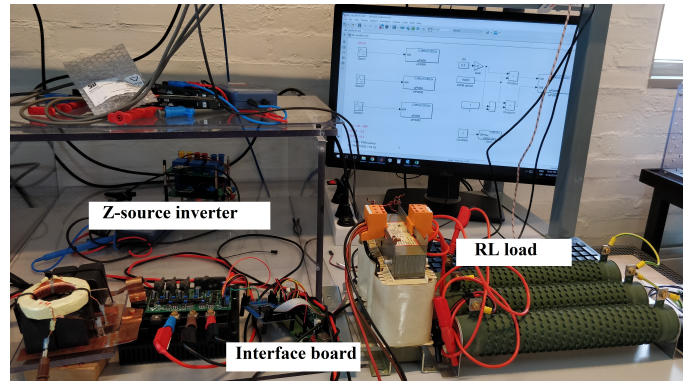


Figure 10. Experimental test bench.

280 5.1. Normal Operation

281 The output current waveform during normal operation of the proposed system are shown in
 282 Figure 11a. It can be seen that ripple-free waveforms are obtained across the three-phase winding of
 283 the motor. It depicts the continuous conduction mode of the proposed system.

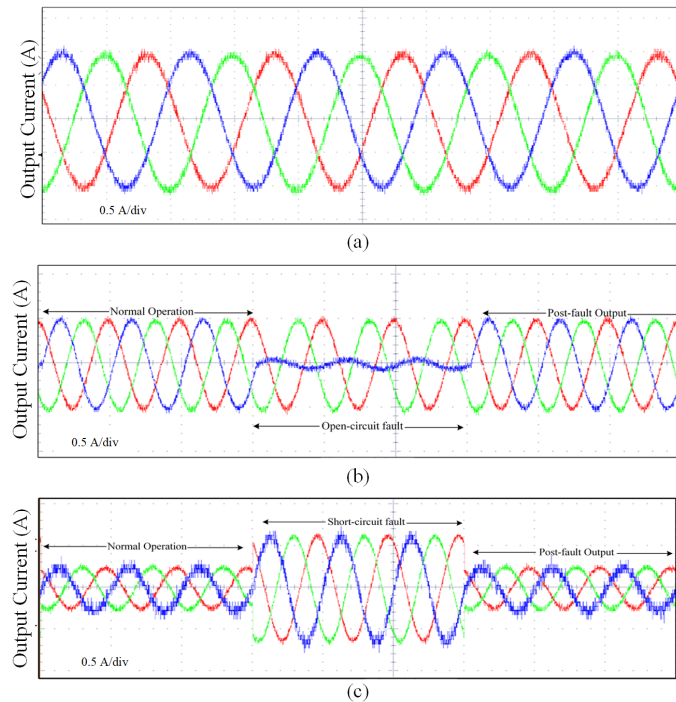


Figure 11. Experimental results: (a) normal operation, (b) open-circuit fault, and (c) short-circuit fault.

284 5.2. Fault-tolerant Operation

285 The output current response of the proposed system is observed for the fault-tolerant operation. To
 286 study the faulty conditions, fault-injection is done in the IGBTs of phase-A of the inverter module. The
 287 output current waveforms during pre-fault, faulty and post-fault operation are shown in Figure 11b
 288 and Figure 11c. It is observed in Figure 11b that due to the open-circuit fault, the output current of the
 289 faulty phase-A is nearly zero. Once the fault-tolerant control is applied, the load current restores back
 290 the motor operation in due time. To investigate the short-circuit fault, the switches in the experimental
 291 setup are kept in ON-state for an infinite period. The output current waveform during pre-fault, faulty
 292 and post-fault operation for the short-circuit fault are shown in Figure 11c. It is observed that the
 293 proposed system restores to the normal operation after diagnosing the fault. The proposed control
 294 scheme provided service continuity at full power in event of faults. The experimental results are

295 similar to the simulation results and confirms the validity of the proposed fault-tolerant strategy for
296 the proposed system.

297 6. Compatibility of Proposed System for Irrigation Application

298 The achieved performance, detailed in Section 4 and 5, show that the proposed system has
299 efficient fault-tolerance capability under switch failures. It is due to the better compensation strategies
300 implemented under faulty operation. However, the compatibility of the proposed system is required
301 to be investigated for the irrigation application. In view of this, the proposed system is analysed for
302 three important factors [43–46]. Firstly, the voltage gain of proposed converter topology is required
303 to be investigated. Secondly, as an essential component for irrigation, the operation of motor is to
304 be regulated in such a way that an optimized system efficiency is achieved without any transients.
305 Thirdly, the motor should exhibit better torque-speed characteristics. In this regard, the characteristics
306 of the proposed system are demonstrated as follows:

307 6.1. Voltage Gain

308 There is a constant and intense demand for reliable, efficient, and small size step-up dc-dc
309 converter in irrigation applications. To cater this issue, the proposed converter had shown continuous
310 input current under variable load conditions. The voltage gain of the proposed converter is high
311 enough to correspond with continuous high voltage applications. The voltage gain by proposed
312 converter-inverter module is 3, as shown in Figure 12. It is inferred that the proposed Z-source-source
313 network configuration provides a high boosting capability. It can also operate with variable and low
314 duty cycle, which simplifies the design procedure and improves the performance. All of the mentioned
315 features made the proposed converter topology a good candidate for irrigation application.

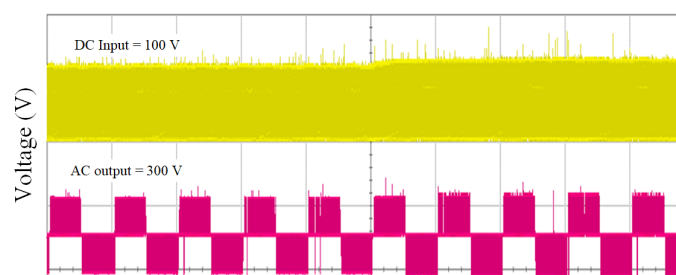


Figure 12. Voltage waveform of proposed converter-inverter topology.

316 6.2. Harmonic Response

317 The efficient response of the proposed system towards transients is indicated by the percentage
318 ripple of the output current supplied to the motor drive. The harmonic spectrum under normal
319 operation is shown in Figure 13a. It is observed that under normal operation the total harmonic
320 distortion (THD) is 0.42% with a high magnitude of the fundamental component.

321 Due to faults, the harmonic distortions are also introduced in the system. These distortions results
322 in high ripples in the load current. Open-circuit faults induce a dc current offset in both normal and
323 faulty phases. This offset current results in a pulsating torque at the stator current and hence in a
324 reduction of the average torque for the drive. Additionally, transients are introduced in the system
325 which deteriorate the power quality of the load current. The harmonic spectrum during this fault is
326 shown in Figure 13b. It can be seen that the THD in case of the open-circuit fault is significantly high,
327 with a magnitude of 44.36 %.

328 The short-circuit currents creates unbalancing in the all phases. The transient response during
329 such fault shows that the load current waveform is highly distorted. The harmonic spectrum during
330 this fault is shown in Figure 13c. The THD in case of a short-circuit fault is significantly high, with
331 magnitude of 25.21 %.

332 Nevertheless, with the proposed fault-tolerant control method, the harmonic spectrum is close to
 333 the value that of normal operation. It can be seen in Figure 13d that the THD in case of FTC operation
 334 is negligible, with magnitude of 1.78 %.

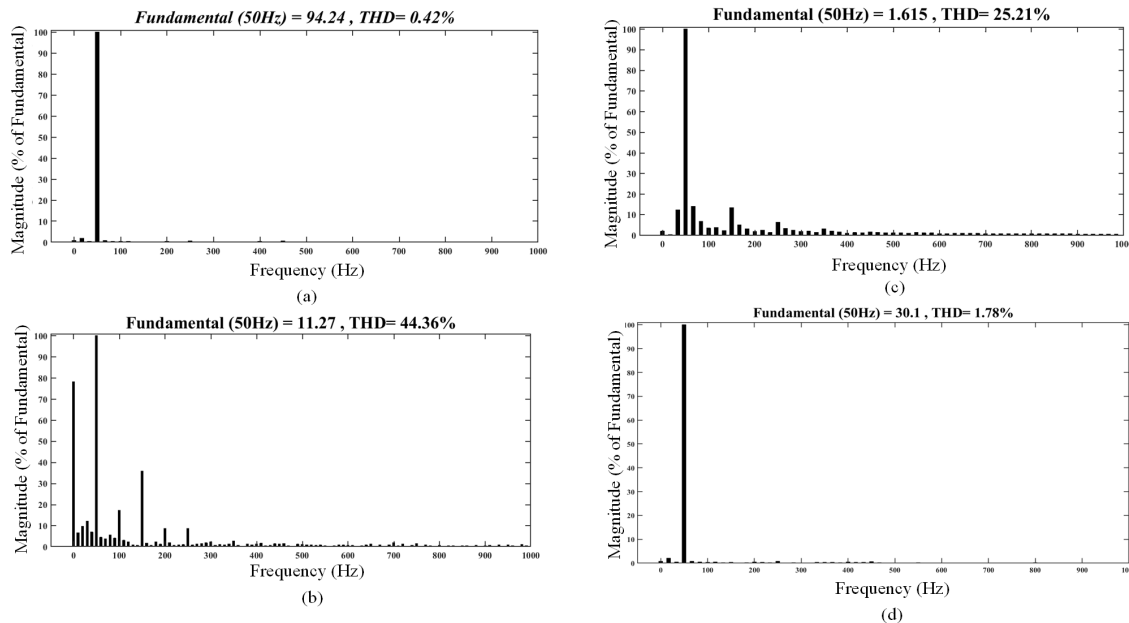


Figure 13. Harmonic spectrum: (a) normal operation, (b) open-circuit fault, (c) short-circuit fault, and (d) FTC operation.

335 Similarly, the harmonic spectrum results under experimental analysis are determined for the load
 336 current of the given system under fault-tolerant operation. The observations of THD for simulation
 337 and experimental analysis are listed in Table 5.

Table 5. Harmonic spectrum results.

Condition	Simulation Results	Experimental Results
Normal operation	0.42	1.23
Open-circuit fault	44.36	41.23
Short-circuit fault	25.21	22.36
Post-fault operation	1.78	1.52

338 It is worth highlighting from Table 5 that the total harmonic distortions values obtained with
 339 the proposed fault-tolerant operation are within the permissible standards IEEE 519-2014 and proves
 340 that this method is suitable in terms of power quality to implement for the motor drives for irrigation
 341 application.

342 6.3. Motor Performance

343 The achieved results demonstrated that the converter topology and FTC operation can compensate
 344 the current and torque with high power quality which results in improved the drive performance in
 345 fault conditions. The obtained results shows that the motor performance maintains a similar operation
 346 in pre-fault and post-fault operation upon implementation of the proposed fault-tolerant scheme.
 347 Also, the rated torque is restored back to the normal value during post-fault operation as shown in
 348 Figure 14a. Similarly, the motor speed, as shown in Figure 14b, gradually increases and reaches to the
 349 rated value with no fluctuations under FTC operation. The achieved torque-speed characteristics are
 350 suitable for irrigation application requiring constant speed operation under variable-load condition.

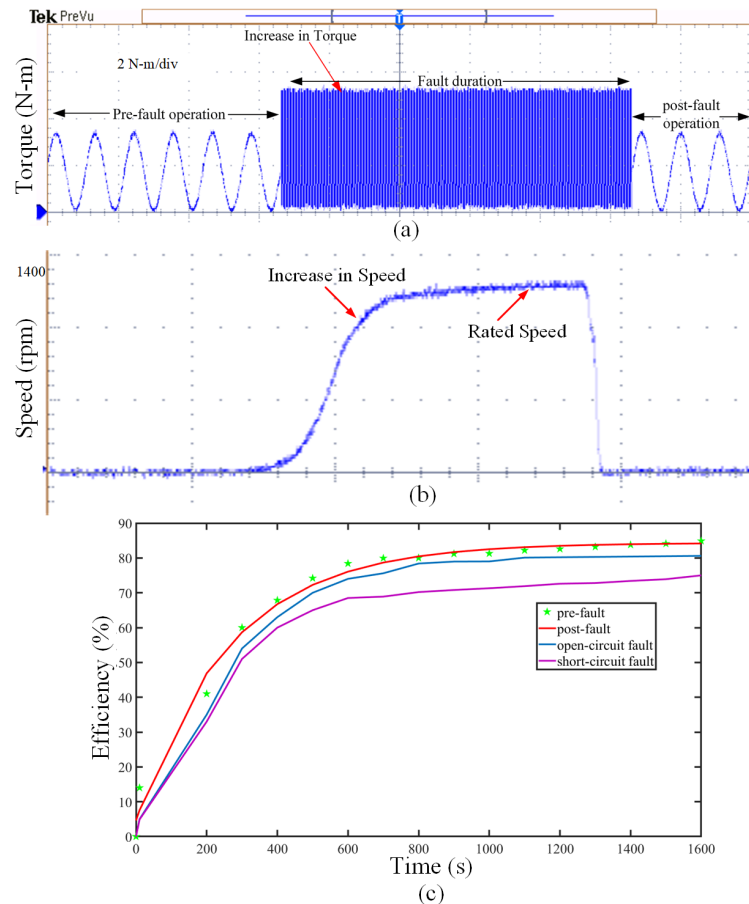


Figure 14. Motor performance under FTC operation: (a) torque response, (b) speed response, and (c) efficiency curve.

351 In addition to it, the motor efficiency is analysed in this section and shown in Figure 14c to validate
 352 the characteristics obtained during the pre-fault and post-fault operation. It can be observed that the
 353 motor efficiency is significantly reduced during open-circuit and short-circuit faults. However, due to
 354 the proposed FTC method the motor efficiency during post fault operation is maintained nearly equal
 355 to value of no-fault operation of the drive. The proposed drive system with fault-tolerant capability
 356 has exhibited an ability to operate at constant power over a wide speed range with excellent overall
 357 performance, and high efficiency.

358 The achieved results show that the system is suitable for the application of the motor-pump
 359 arrangement for rural irrigation and is less prone to shutdown with switch failures.

360 7. Conclusions

361 In this paper, implementation of a PV-and ZSI-fed motor drive system with fault-tolerant
 362 capability for rural irrigation is designed and implemented. The proposed modulation scheme for
 363 Z-source inverter enables it to obtain a high voltage gain with a minimum component requirement.
 364 Additionally, a suitably fast and efficient fault-tolerant scheme is presented for the proposed system. It
 365 operate continuously without reduced output power and can be applied for all kinds of motor-drive
 366 applications. The reliability of the proposed topology is validated by simulation and experimental
 367 results. The proposed control scheme can eliminate the harmonic distortions issues related to the
 368 load current during the faulty operation. The motor performance parameters such as torque, speed,
 369 harmonic spectrum and efficiency curve are presented to prove the system's compatibility for irrigation
 370 application. The proposed system with a fault-tolerant capability proves to be an efficient and
 371 economical solution to the irrigation problems of the rural areas in developing countries.

372 **Author Contributions:** V.S. conceived, designed and performed the simulation and experimental evaluations and
 373 was responsible for preparing the original draft of this paper; J.H. supervised for the development of the power
 374 topology and conceptual solution and is responsible for reviewing and editing of this paper; S.M.N.A and M.K.
 375 were responsible for the guidance towards relevant key theoretical and technical suggestions.

376 **Conflicts of Interest:** The authors declare no conflict of interest.

377 Abbreviations

378 The following abbreviations are used in this manuscript:

379	PWM	Pulse width modulation
	IGBT	Insulated gate bipolar transistor
	THD	Total harmonic distortion
	PV	Photo-voltaic
380	ZSI	Impedance source inverter
	VSI	Voltage source inverter
	FTC	Fault-tolerant Control
	CCM	Continuous conduction mode
	IC	Integrated Circuit
	MPP	Maximum Power Point

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